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Peak to Average Power Ratio Reduction for Filtered OFDM Modulation in 5G Frameworks Using Hybrid Techniques

M. Selvakumar*, B. Sudhakar
Assistant Professor, Department of Electronics and Communication Engineering, Government College of Engineering,Sengipatti, Thanjavur, Tamil Nadu, India.
Department of Electronics and Communication Engineering, Faculty of Engineering and Technology, Annamalai University, Tamil Nadu, India.
**Email: ***ms.lecturer@gmail.com, balrajsudhakar@gmail.com

Abstract: Filtered Orthognal Frequency Division Multiplexing (FOFDM) is the upcoming modulation structure for 5G communication framework with higher spectral efficiency as it has lower side-lobes. However, in all multicarrier modulations, it aches from abnormal peaktoaverage power ratio (PAPR) thatshakes the power amplifiers linearitywhichclues in connection with signal distortion and spectral regrowth with abnormal Bit error rate. In the system, it lowers power efficiency and poor battery life is due to high PAPR, which encumbers in the implementation of Multicarrier modulation.In the literature a more number of single techniques for peak to average power reduction were engaged but results in relatively low reduction, also it has some failingsthat may influence the presentation of framework boundaries like Bit Error Rate and spectral efficiency. Therefore, a combined technique is presented, i.e. precoding with companding for more reduction in the PAPR to maintain the overall system performance in the 5g networks.The precoding method spreads the energy f data symbol over the subcarriers which minimized the autocorrelation coefficients of IFFT input that lessen the PAPR of OFDM signal into the single carrier level transmission.The statistical characteristics level of the multicarrier outcome signals by the Rayleigh distribution into uniform distribution which results in peak and normal power levels of OFDM signal due to the companding process and consequently reduces the PAPR.The proposed techniques combine the precoding techniques: Walsh Hadamard Transform (WHT) and Discrete Hartley Transform (DHT) with two novel non-linear companding techniques: Absolute Exponential Companding (AEC) and Rooting Companding Technique (RCT). We have examined, compared, and analyzed the peak average power ratio for filtered OFDM using both conventional single and hybrid techniques. The companding techniques have shown a higher reduction in PAPR when compared with the precoding techniques. The hybrid Discrete Hartley Transform with Root Companding Technique has shown the least PAPR of 4.7dB and the highest reduction performance in all used techniques with a reduction of about 6.1 less thanReal FOFDM peaktoaverage power ratio. From the outcomes, the hybrid procedures have indicated a superior execution by adequately diminishing the peak to average power proportion without corrupting the presentation of framework when contrasted and the traditional single techniques.

Keywords: 5G, OFDM, PAPR Reduction, Digital modulation.

1. Introduction

The up-and-coming age of 5^{th} generation mobile communication (5G) will offer something other than the customary Mobile Broadband Band services. The fundamental applications will give are isolated into three significant classes such as enhanced Mobile BroadBand (eMBB), massive Machine Type Communication (mMTC), and ultra-Reliable and Low Latency Communication (uRLLC). Hence, to satisfy the prerequisite of these applications many research works are going on in various extensions. New modulation techniques were developed and suggested to be alternative to the OFDM modulation technique. Orthogonal Fre-

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quency Division Multiplexing (OFDM) utilizes a square window in the time domain permitting an extremely effective usage which was embraced as air interface in a few wireless communication standards, together with the 3rd generation organization (3GPP), Long Term Evolution (LTE), and IEEE 802.11 standard families because of related promotional merits such as robustness besides multipath fading, simple execution, proficient one-tap frequency domain equalization empowered by utilizing the CyclicPrefix (CP), and uncomplicated along with simple augmentation to enormous Multiple Input MultipleOutput (MIMO) and elevated beam gain framing solutions.Regardless of its benefits, OFDM experiences various disadvantages including high PAPR and lobes of high side in frequency. OFDM needs tough time synchronization to keep up the symmetry among various User Equipment (UE).Therefore, signaling overhead increments of the quantity of UEs in an OFDM-oriented framework. Apart from this OFDM has huge sensitivity with respect to the Carrier Frequency Offset (CFO) disparity among the diverse devices [1].

1.1. Characteristics of 5G systems

All those drawbacks restrict the approval of OFDM in the 5G air interface to get the following key characteristics envisioned for 5G wireless networks at present [1].

- 1000 x high volume mobile data per topographicalregion
- 10–100 x additionallinked components.
- 10–100 x average data rate for client is high.
- 10 x less energy utilization.
- End-to-end latency with <1 ms
- Worldwide 5G access even in areas of low density areas.

1.2. Scenarios of 5G Systems

The fundamental characteristics mentioned in the previous section are visualized in connection with scenarios specified belowfor 5G research community as in [2]. They are bit pipe communication. Internet of things (IoT), Tactile internet, and Wireless Regional Area Networks (WRAN). The prerequisites of these various situations can affectwaveform selection. Along these lines, to mention the downsides of OFDM and empower previously mentioned qualities, diverse physical layer waveforms are researched for 5thGeneration network.There are several classifications of MCM candidate schemes for 5G. In [3] the proposed waveforms are classified into four main categories:OFDM based Waveforms for DFT Spread,Single Carrier based Waveforms, Non-Orthogonal Waveforms, and OFDM reliantWaveforms.

F-OFDM is uniqueof the OFDM oriented competitor modulations for forthcoming 5thGeneration frameworks. Nonetheless, all MCM have high PAPR affect significant downside of this regulation, however,transmitted signals are the superposition of different autonomous with symmetrical subcarriers. Lesser PAPR is basic waveform plan standard for the 5th generation new waveforms to be skilled to meet the 5G prerequisites and to improve efficien-

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cy of power, so PAPR ought to be decreased.It has gotten basic to utilize productive PAPR decrease methods, but employing single techniques results in relatively low reduction of PAPR, also it may have some demerits which can shake the performance of other system constraints like BER and spectral efficiency. So hybrid techniques are considered to be a better choice as it exploits the advantages of both techniques used in hybridization and will efficientlylessen the peak-to-average power ratio without generating spectrum side-lobes or degrading bit error rate routine of the frameworks.

The following content is coordinated as follows. Section two educate FOFDM. PAPR in section three, section four shows the proposed research work. The outcomes are explained in section five and section six shows the conclusion.

2. Filtered Orthogonal Frequency Division Multiplexing

The bandwidth assigned in FOFDM is further be fragmented intoseveralsubbands; every subband is alone modulated using OFDM modulation [4], Sub-band-established filtering, which further applied to overwhelm the intersubband interference, F-OFDM moreover provides majorlessenover the guard band consumption, that leads to well-organized spectrum utilization [5]. Further, extension of classic OFDM with additional subband filter along with flexibility in varying the parameters such as transmission time duration, cyclic prefix length along with subcarrier spacing [6]. Filtered bandwidthis designed as a certain sub-band but it’s not essentially equivalent to 1 PRB. In this fashion, F-OFDM has the ability to resolve the restrictions of the OFDM and also sustaining the merits. While important assets of F-OFDM are below [5], [7]:

a. In subband trivet filtering, worldwide synchronization necessity loses, it could be up holding inter-subband asynchronous communication.
b. The guard band consumption could be reduced to a minimal level, by creating filters to suppress OOB emission.
c. An enhancednumerical technique could be applied in each subband toward ensemble demands of specific kinds of services.
d. Core waveform of OFDMthat allows F-OFDM to like theirnecessarybelongings of OFDM and OFDM-based designs are assisted for instant application of previous.
e. F-OFDM comes compatible with MIMO and their PAPR could be easily lessen using conventionalreduction techniques.
f. F-OFDM empowers the conjunction of various time-frequency granularities, e.g., diversesub-carrier spacing's, in the framework so various pieces of waveform which frequency could be enhanced for various application and transmission conditions, whilethe capacity, it is a critical element of product characterized airborne interface.

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The primarydiverse between the F-OFDM, their classical pulse-shaped OFDM is that the traditional pulseshaped OFDM achieves pulse shaping per subcarrier by smearing time-windowing along withevery CP-OFDM symbol towardenhanced spectrum localization. Conversely, F-OFDM accomplishes well spectrum localization by stifling their out-of-band outflow. Double procedures could coincide, i.e., F-OFDM could be used on pulse shaped molded CP-OFDM to additionally enhance their presentation [7].

2.1. Filtered-Orthogonal Frequency DivisionMultiplexingTransceiver

As delineated in Fig.1, where F-OFDM transmitting device produces its OFDM signal dependent on assigned square of consecutive sub-carriers of various continuous OFDM symbols.

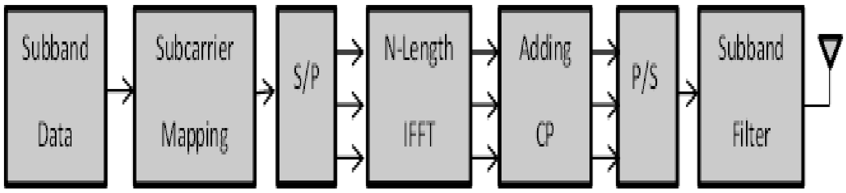

Fig.1. FOFDM Sender

To comprehend the sender and tuner side cycles, they need to describe the F-OFDM symbol numerically. Beginning there to discrete-time portrayal for OFDM symbol:

$$x(n) = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} c_k e^{j2\pi kn/N} \tag{1}$$

Since N is the subcarrier count. F-OFDM signal is furtherattained by feeding their signal *x(n)*over a suitablyplanned spectrum shaping filter toward suppressing only their out of band emissions, it is obtained as:

$$\tilde{x}(n) = x(n) * f(n) \tag{2}$$

Where spectrum influencing filter*f (n)*come balanced along with frequency at theirallotted subcarriers, which bandwidth areproportion to total frequency width with the allotted subcarriers, and OFDM symbol length is segmented.

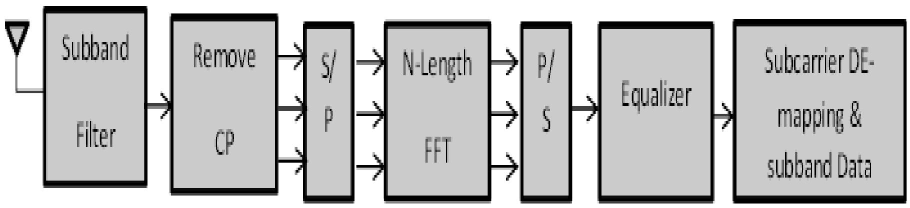

Fig.2. FOFDM Receiver

2.2. Previous works

Several works have been made to determine the considerable enhance the OFDM spectral competence and thus fulfill the necessities of 5G applications. [8] presented an asynchronous

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F-OFDMA for removing the side lobe leakages and destroy the synchronous signalling of OFDM. The BER results are investigated and power spectral density (PSD) of the design presented are compared to the UFMC waveform candidate. It is noted that the presented design is effective compared to the UFMC candidate based on the frequency localization and BER performance. [9] projected an extended version of F-OFDM waveform with maximum tractability. It is obtained by the division of the bandwidth to different subbands and employing filtering for all bands to attain distinct service kinds. [10] projected a study of the F-OFDM depending upon the spectrum slicing in the real 5G background. The presented model has achieved maximum spectrum efficiency for the asynchronous application in the expected 5G network. [11] has explained the spectral efficacy outcome of F-OFDM in a 5G area test. The resultant values ensured that the F-OFDM waveform model has resulted in better spectral efficiency over the LTE-OFDM model. Liu and others [12] carried out a comparative analysis of the filtering based waveform candidate. The author has explained the fullband filtering candidate over other waveform candidates with respect to OOBE leakage. [13] intended to explain the problem of PAPR to minimize the peak value of the OFDM signal.

3. PAPR

Every OFDM-founded system has maximum PAPR as thatconveyed signal is superposition of many subcarriers via an IFFT operation. Similarly,F-OFDM signals moreoverencompass of numerous autonomous and symmetrical subcarrier and suffers from the problem of maximum PAPR that effectively uniqueof important implementation disadvantages of OFDM-founded systems. Also, PAPR can be expanded with subband c in F-OFDM framework.

3.1. Papr Definition

The PAPR of the time baseband OFDM signal x (t) is characterized as the proportion among the higher quick power to their normal power, which is defined below:

$$PAPR\big(x(t)\big) \triangleq \frac{\max_{0\le t\le Kt_s} |x(t)|^2}{\frac{1}{Kt_s}\int_0^{Kts} E\{|x(t)|^2\}dt} \tag{3}$$

Whereas E [·] representspredictable value. When x(t) signal is sampling at the rate of Nyquist $t = nt_s$, along with number n, so we can write the discretetime baseband OFDM signal x(n) as:

$$x(n) = \frac{1}{\sqrt{K}}\sum_{k=0}^{k-1} X(k)e^{j2\pi\frac{k}{K}n}, \quad n = 0,1, \dots K-1, \tag{4}$$

Furthermore, the PAPR regarding discrete-time baseband OFDM signal couldstate as:

$$PAPR(x(n)) \triangleq \frac{\max_{0\le n\le K-1} |x(n)|^2}{\frac{1}{K}\sum_{n=0}^{k-1}|x(n)|^2} \tag{5}$$

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Inutmostscenarios, the PAPR of discrete OFDM signal is smaller than the PAPR of continuous OFDM signals by 0.5 ~ 1 dB [14]. Hence, relationship within PAPR is rendered by:

$$PAPR(x(n)) \leq PAPR(x(t)) \tag{6}$$

The above power credits can similarly be portrayed with respect to their sizes (not power) by describing the peak factor (CF), which is described as the extent between most extraordinary magnitude of OFDM signal x (t) and rootmeansquare (RMS) of waveform [3]. CF is defined by:

$$CF = \frac{\max\ |x(t)|}{E[||x(t)|^2|]} = \sqrt{PAPR} \tag{7}$$

PAPR is a discretionary variable since it is a component of input data, along with input data is an irregular variable.Hence PAPR can be determined by utilizing level intersection rate hypothesis that ascertains the normal number of times that the envelope of a sign crosses a given level. Knowing the amplitude distribution of the OFDM yield signals, it is anything but difficult to register the likelihood that the instantaneous amplitude will be over a given threshold and the equivalent goes for power. The interpretation is computing with complementary cumulative distribution function (CCDF) utilize by condition (7) for different PAPR values as render:

$$CCDF = \Pr\,(PAPR > PAPR0) \tag{8}$$

The Cumulative Distribution Function (CDF), one of most concordantutilize boundaries, is used through quantifying the proficiency of any PAPR method. The CCDF causes us to quantify the possibility that PAPR exact information block exceeds the given threshold [15].

3.2. Precoding Models

The intention of precoding strategies to acquire a signal with lesser PAPR rather traditional OFDM that absence of precoding procedures to lessen the severalusers' interference. The PAPR decreasing must remunerate the nonlinearity of the HPA having as impact the decrease of the bit error rate.A few of the important precoding techniques are TheDiscrete Sine Transform (DST), Zadoff-chu Transform (ZCT), Discrete Cosine Transform (DCT), and Discrete Fourier Transform (DFT) [16].

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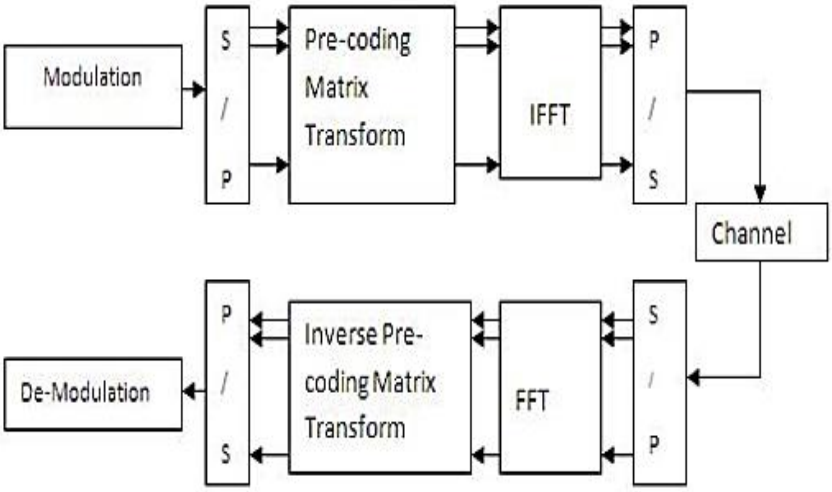

Fig.3. Block Diagram of Precoding Method for PAPR Lessening

As specified in Fig.4.7 above, a precoding technique comprise of multiplying the modulated data of every OFDM slab through a precoding matrix ahead of subset assignment IDFT block. In [17], While MC symbol with N symmetrical subcarriers, assume an input vector with N symbols is depicted as

$$\mathrm{X} = [\mathrm{X}_{\mathbf{0}},\mathrm{X}_{\mathbf{1}},\mathrm{X}_{\mathbf{2}}, \ldots \ldots \mathrm{X}_{\boldsymbol{N-1}},]^{T}, \quad (9)$$

The precoding plan that utilizes the precoder to diminish the PAPR is assumed as a preprocesswhich is employed amongst the serial to parallel adaptationand the IFFT process as in Fig 3. The precoding matrix can be a similar size as the allocated subcarrier vector. The measurementof precoding matrix, denoted by P, is, hence, N × N. It is used as a spreading code and can be expressed as:

$$\begin{bmatrix} P_{0,0} & P_{0,1} & \Lambda & P_{0(,N-1)} \\ P_{1,0} & P_{1,1} & \Lambda & P_{1,(N-1)} \\ \mathrm{M} & \mathrm{M} & \mathrm{O} & \mathrm{M} \\ P_{(N-1),0} & P_{(N-1),1} & \Lambda & P_{(N-1),(N-1)} \end{bmatrix} \quad (10)$$

Next input vector X is multiplied with the precoding matrix NXN, the output vector Y has a distance N described such

$$\mathrm{Y} = \mathrm{P} \otimes \mathrm{X} = [\mathrm{Y}_{\mathbf{0}},\mathrm{Y}_{\mathbf{1}},\mathrm{Y}_{\mathbf{2}}, \ldots \ldots \mathrm{Y}_{\boldsymbol{N-1}},]^{T}, \quad (11)$$

Where P signifies the precoding matrix in (), and ⊗signifies element-wise multiplication.To shorten rear precoding progression, the P matrix has to be a unitary matrix which means PP^{H}=I_N. Also, as the precoding is performed before the IDFT, it should contain a signal transform to the frequency domain. Hence, we obtain:

$$P = [\mathrm{P_{ij}}]_{\mathrm{i,j} \in [\![0,\mathrm{N}-1]\!]} \quad (12)$$

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In N × N where is P is precoding Matrix is shown in above equation.The complex baseband FOFDM signal with N subcarriers is defined by:

$$x(n) = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} P.c_k e^{\frac{j2\pi kn}{N}} \tag{13}$$

Where 0 ≤ t ≤ NT,The modulated FOFDM vector signals among N the subcarriers as follows: $X_N = IFFT\{P.X_N\}$

3.3. Companding

Compressing and Expanding denoted as Companding which a PAPR reduction technique for MC signal, established on speech processing procedure μ−law. In order to achieve the companding methodin MC signal, which is companded and quantized earlier than transformed into an analog waveform. At receiver, the digital form is converted from the received signal first and then expanded.

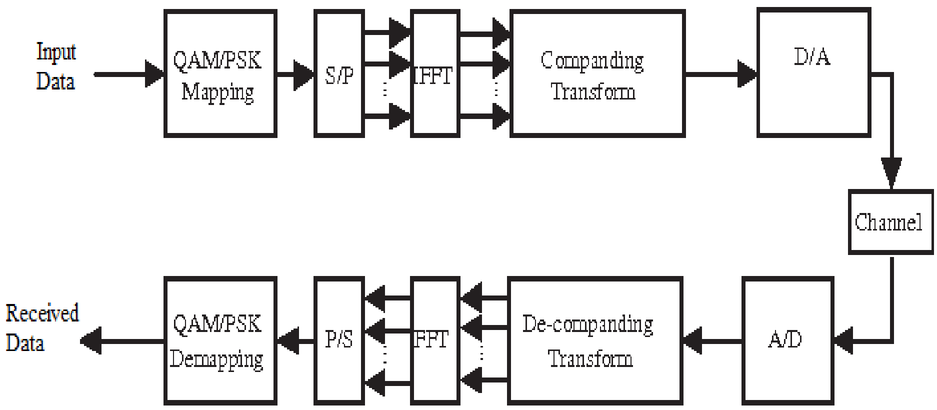

Fig.4. Block Diagram of CompandingMethod for PAPR Lessening

Non-linear companding is a particular cut-out procedure which offers great PAPR decrease with better BER execution, low usage multifaceted nature, and no bandwidth capacity development. The companding transform compand the first signals utilizing severe monotone expanding capacity; accordingly, the companded signals can be recouped accurately through the comparing reversal of companding change at the recipient. Signals don't influence small signals, though companding expands the little signals while packing the huge amplitude signals. A ton of companding methods is available. The essential idea of the vast majority of the companding methods thereto changes their Rayleigh appropriated OFDM signal toward a consistently conveyed signal by packing the signal at the information and growing the signal at yield to keep the sign level over the commotion level during handling. At the output, the first info signal is then reestablished by a simple attenuation. Companding builds the SNR, where information signal is less and accordingly diminishing the impact of a framework's noise source.

We utilized the compander to decrease PAPR of MC signal on the grounds that compander enhances the little sign and increments the normal intensity of the MC signal and lessens the

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high peak value. Notwithstanding, a compander comprises a blower at the sender side and an expander at the beneficiary side. Consequently, toward finish of the transmitter side, the pressure will occur; the development cycle will happen at the receiver side [18-21].

4. Proposed Methods

FOFDM is unique method of OFDM related applicant modulations for forthcoming Fifth-generation frameworks. Be that as it may, be all MCM have high PAPR is the significant disadvantage of this modulations as the sent signals are superposition of various free plus symmetrical subcarriers. Lesser PAPR is a basic waveform plan guideline for the 5g new waveforms through skilled to meet the 5G necessities in addition to improving power effectiveness, so PAPR ought to be decreased.It has become authoritative to employ effective PAPR lessening techniques, but employing single techniques results in relatively low reduction of PAPR, also it may have some demerits which may shake performance of other system constraints like BER and spectral efficiency. So, hybrid techniques are considered to be a better choice as it exploits the merits of both techniques used in hybridization and will efficiently-lessen the peak-to-average power ratio without generating spectrum side-lobes or debasing bit error rate recital of system.So, here suggested a crossbreed technique that combines Walsh-Hadamard Transform and Discrete Hartley Transform precoder techniques with two novel non-linear companding, Absolute Exponential Companding (AEC) and Rooting Companding Techniques (RCT) which based on Exponential Companding and Square-Rooting Companding respectively for PAPR reduction of Filtered-OFDM modulation. Our goal is to take the merits of both techniques for additionallessening in the PAPR of the system.

The precoding techniques depend on the re-direction or spreading the energy of information image over the subcarriers designated to the client prior to taking IFFT which can diminish the PAPR of OFDM signal of single-carrier transmission to be even. Furthermore, using precoding can improve PAPR without escalating the system complexity or demolish the orthogonality between the subcarriers. Also, the precoding technique develops the BER performance.The statistical characteristics of the multicarrier output signals are changed by companding process through Rayleigh distribution to uniform distribution that will result in changes in both the power levels of OFDM signals and accordingly, decreases considerably the PAPR. Unlike the basic μ-law and a-law companding techniques, the proposed companding techniques guarantee the companded signals will not be smaller than expanded signals that were originally small.

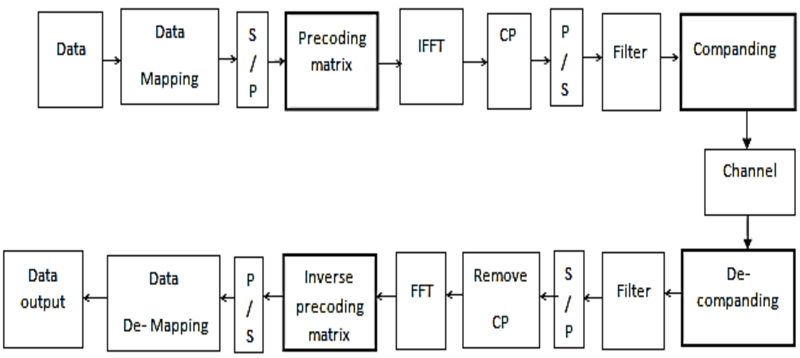

Fig. 5. Block Diagram of the Proposed Hybrid Techniques for PAPR Reduction of filtered OFDM

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Thesuggested hybrid PAPR lessening scheme applied to the F-OFDM modulation combines three precoding techniques with two novel companding techniques. During the transmission process,the modulated data stream is primarilyconverted by predefined precoding matrix before applying IFFT which helps in minimizing the autocorrelation coefficients of IFFT input, then data is companded bycompressing and enlarging of large amplitudes and small amplitudes respectively without affecting average amplitudes and hence, further reduction of PAPR.At the receiver end,the reverse operation of both companding and precoding is carried out followed by an equalizer and data detection. The PAPR will be assessed statistically by means of Complementary Cumulative Distribution Function. Finally,analyzein addition to select best pair of hybrid techniques.

4.1. Walsh-Hadmard Transform (WHT) Precoding

The Hadamard transform (besides identified as the WalshHadamard transform, Hadamard Rademacher–Walsh transform) is one among the comprehensivecourse of Fourier transforms. Hadamard transform achieves an involution, symmetric and orthogonal along with linear operation on real numbers. Hadamard transform is being composed of size 2×2×...…×2×2 DFTs, and the familiar fact is it can be equated to a multidimensional DFT size. Hadamard transform is reversibly destructed as an arbitrary input vector towards a position of Walsh functions.WHT stands non-sinusoidal and it is a symmetrical method which disintegrates a sign into set of fundamental functions. These capacities are called Walsh capacities, the hadamard change plot diminishes the event of the high pinnacles contrasting the ordinary OFDM framework. The hadamard change is utilized on the grounds that it diminishes the autocorrelation of the info arrangement to lessen the PAPR of OFDM subcarrier signals. It also doesn’t need sending side information to the receiver.WHT could be actualized through a butterfly structure in place of FFT. This implies applying WHT doesn't need broad increment thereof framework unpredictability. During the portion of WHT could be composed as complied: -

$$H_1 = [1] \tag{14}$$

$$H_2 = \frac{1}{\sqrt{2}}\begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} \tag{15}$$

$$H_{2N}\frac{1}{\sqrt{2}} = \begin{bmatrix} H_N & H_N \\ H_N & -H_N \end{bmatrix} \tag{16}$$

Where${H_N}^{-1}$ denotes the binary complement of H^N.

Hadamard transform is noted as symmetrical linear transform and could be actualized through a butterfly structure in place of FFT. This implies applying Hadamard transform doesn't need the broad increment of framework intricacy. Though received vector signal x_N corrupted by noise vector n can be recovered to X_N as:

$$X_N = FFT\{H_N^{-1}.x_N\} \tag{17}$$

$$= FFT\{{H_N}^{-1}.IFFT(H_N.X_N) + n) \tag{18}$$

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The Length N of a signal FWHT is notedas:

$$y_n = \frac{1}{N}\sum_{i=0}^{N-1} x_i\, WAL(n.i) \tag{19}$$

$$x_i = \frac{1}{N}\sum_{i=0}^{N-1} y_n\, WAL(n.i) \tag{20}$$

Where i = 0,1... N-1 in addition WAL (n,i) are Walsh functions

4.2. Discrete Hartley Transform (DHT) Precoding

The DHT is a linear, invertible function H: $R^n -> R^n$ (where R represents set of real numbers). In DHT, N real numbers are$X_0, X_1, X_2 \ldots. X_{N-1}$ transformed keen on N real numbers$H_0, H_1, H_2 \ldots H_{N-1}$. The N-point DHT could be described as follows:

$$H_K = \sum_{n=0}^{N-1} X_n \left[\cos \frac{2\pi nk}{N} + \sin \frac{2\pi nk}{N}\right] = \sum_{n=0}^{N-1} x(n) cas \frac{2\pi nk}{N} \tag{21}$$

Where$cas\ \theta = cos\theta + \sin\theta$and $K = 0, 1 \ldots N-1$

$P_{m,n} = cas\left(\frac{2\pi mn}{N}\right)$,m and n are integers from 0 to N-1. P is the pre-coding matrix of size N × N.The DHT is likewise invertible change which permits us to recuperate the X_n from H_k and backward can be obtained by multiplying DHT of H_k by 1/N.

4.3. Rooting Companding Technique (RCT)

We have proposed a modified Square rooting companding (SQR) Technique named the Rooting Companding Technique (RCT) to further increase in PAPR reduction of FOFDM system. This technique is an adaptive square rooting companding Technique based on altering the square rooting to a ratio *r*, we can write the RCT equation as

$$\text{f[x]} = (|x|)^r * \text{sgn(x)} \tag{22}$$

Whence 0.1$\leq r \leq$ 0.9, sgn(x) = sign(x), and it is applied in RCT in order to sustain theirstages of OFDM signal while only changing only the amplitudes. The amplitude level changes depending on the value of *r*. At the beneficiary end the converse of companding capacity is applied to recuperate the yield signal by utilizing the Rooting decompanding equation:

$$f[k]^{-1} = |x|^{\frac{1}{R}} * sgn(x) \tag{23}$$

4.4. Absolute Exponential Companding (AEC)

The proposed Absolute Exponential Companding technique is based on the concept of the exponential companding technique and its equation is resulting from combining both trapezoidal power companding along with exponential companding:

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$$\mathrm{h(x)} = \quad \mathrm{sgn\ (x)} \sqrt[d]{\alpha\left[1-\exp\left(-\frac{|x|^2}{\sigma^2}\right)\right]} \qquad (24)$$

By taking the total estimation of the square root parts to kill any further potential phase distortion as the received signal with outstanding companding is exceptionally misshaped that the square root part isan imaginary or complex number.

Whencesgn(x) is sign capacity and the positive consistent α decides the average power output signals. To keep the input and output signals at a similar average power level

$$\alpha = \left[\frac{E[|x|^2]}{E\left[\sqrt[d]{[1-\exp\left(-\frac{|x|^2}{\sigma^2}\right)}\right]}\right]^{\frac{d}{2}} \qquad (25)$$

The reverse capacity of h(x) as the Decompanding activity at the receiver end and it's given by:

$$h^{-1}(x) = \mathrm{sgn\ (x)} \left|\sqrt{-\sigma^2 \log_e\left(1-\frac{|\mathrm{x}|^{\frac{2}{d}}}{\alpha}\right)}\right| \qquad (26)$$

5. Results and Discussion

We plan an F-OFDM transceiver subject to LTE standard as recorded in Table I beneath. Furthermore, we will review the presentation of the filter OFDM with respect topower spectral density, peaktoaverage power ratio in addition bit error rate.

Table 1 Parameter Setting

Parameter	**Value**
FFT Point in Numbers	1024
Data subcarrierNumbers in subband	120
Subcarrier gaps(KHz)	15
Number of symbols	1000
Cyclic prefix length (samples)	32
Channel Bandwidth	10MHz
Modulation	64QAM

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Sampling frequency	30.72MHz
Channel type	AWGN
Filter model	FIR
Filter length	513

In F-OFDM, their time space subband OFDM signal is gone over a very much planned channel to improve the out-of-band radiation of theirsubband signal. Their channel's passband relates to the signalbroadcast capacity, just the couple of subcarriers near their edge are prejudiced. Image obstruction amongcaused is limited because of channel config, utilizing windowing with subtle truncation. There of the force otherworldly densities for F-OFDM are shown in Fig. 6, it shows that F-OFDM has minorside lobeswhich permit a greater usageof dispensed range, prompting expanded otherworldly proficiency.

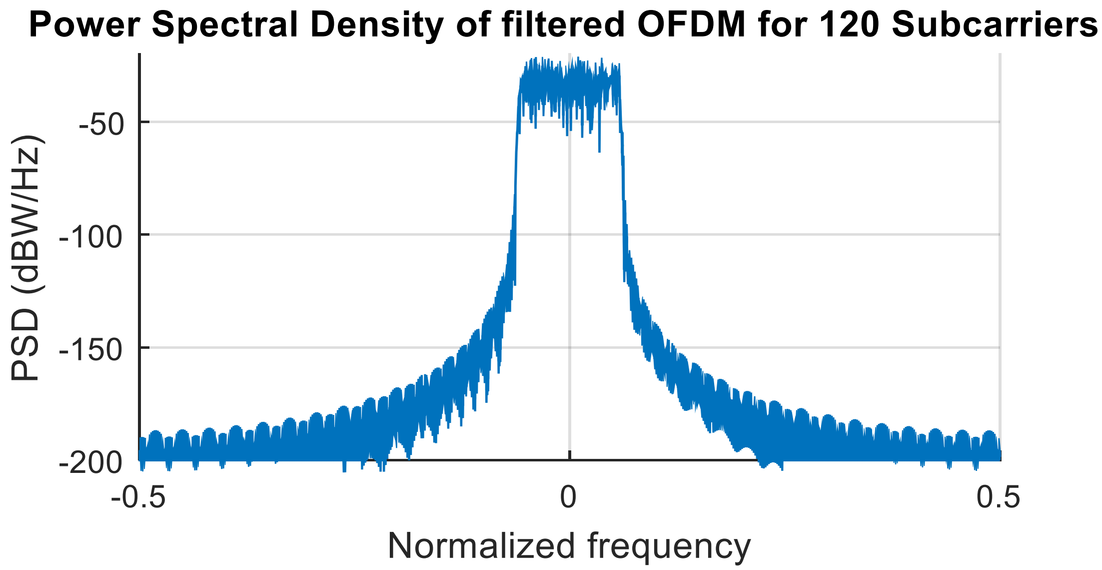

Fig.6. Power Spectral Density of F-OFDM signal

We examine the Peak average power ratio of the filtered OFDM before any lessening techniques to be compared with proposed hybrid PAPR techniques. As depicted in fig.7, the original PAPR for filtered OFDM is 10.8dB that shows a high PAPR as in all multicarrier systems that may limit the FOFDM that is applied in fifth-generationframeworks, whichcompulsory to lessen the PAPR.

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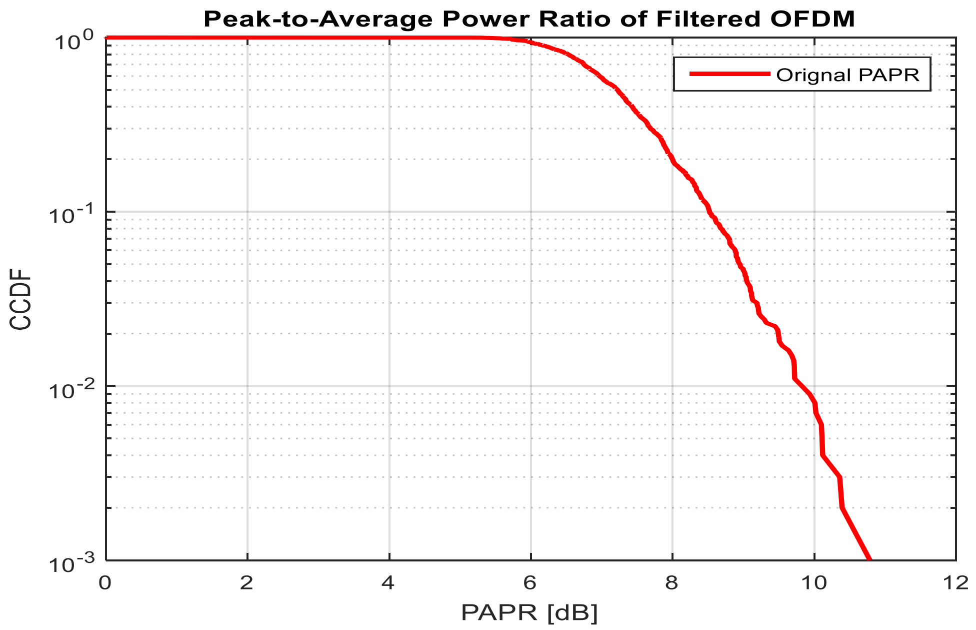

Fig.7. Peak toAverage Power Ratio for Filtered OFDM

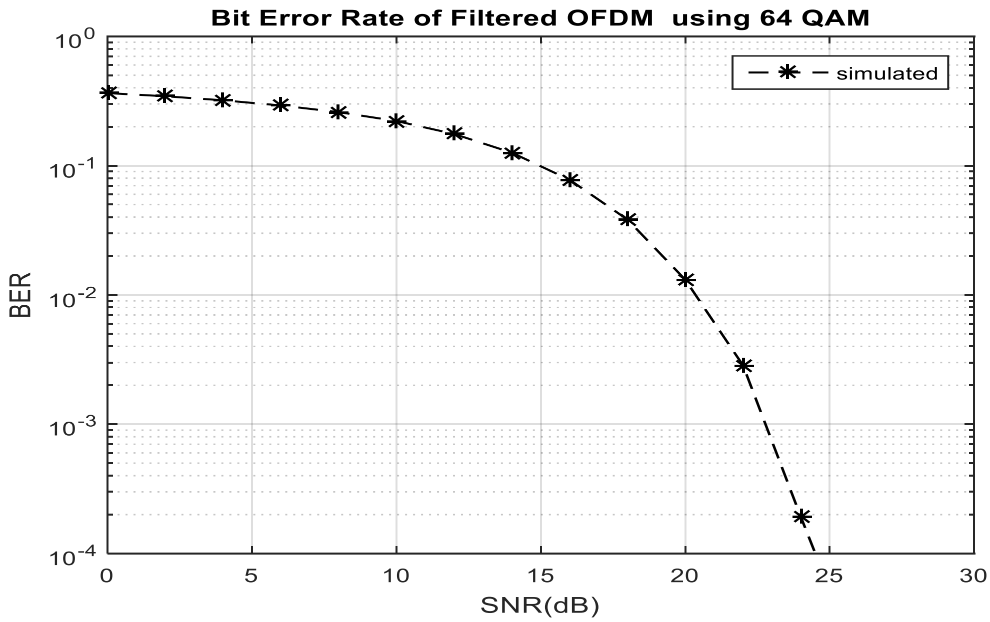

Fig.8. Bit Error Rate of Filtered OFDM

Fig. 8 Shows the Bit Errorchart of F-OFDM system. It shows that Filtered-OFDM system has a relatively high BER.Before applying the hybrid PAPR reduction techniques, we first applied each of the companding and precoding techniques to the system and then compared the performance of each pair of the techniques as shown in fig. 9&10 for precoding and companding techniques correspondingly.Figure 9illustrates the contrast in PAPR reduction performance between DHT and WHT precoding techniques with the original PAPR. DHT has shown a much improvement than WHT with a PAPR of 7.2dB whereas the PAPR of WHT is 9.7dB.On the other hand, in figure 10 we have compared both RCT and AEC companding techniques with original PAPR; we have observed a reduction in PAPR with 6.8dB and

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6.5dB for RCT and AEC respectively. As per the result, the RCT has shown a slight improvement in PAPR reduction than that of AEC.

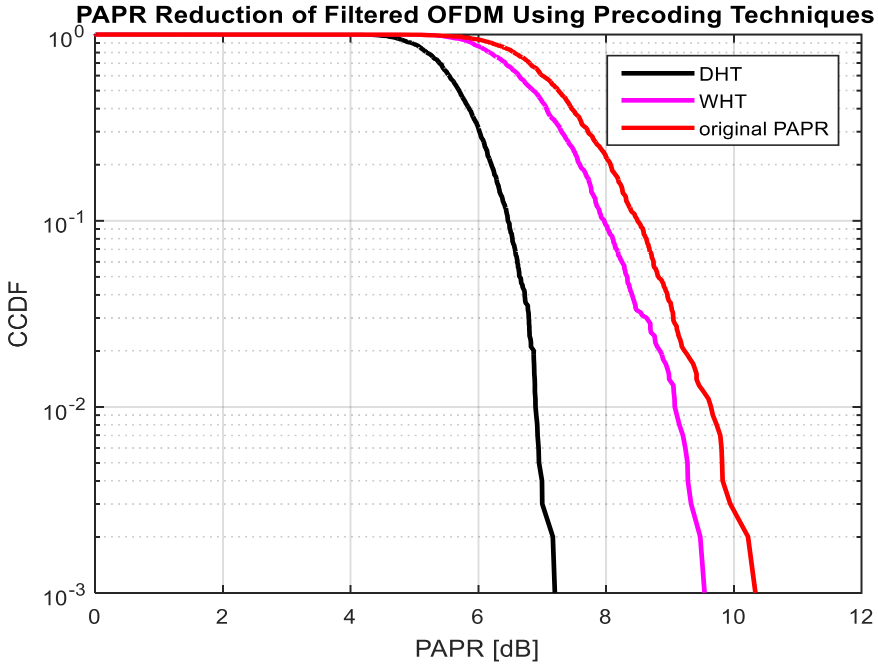

Fig.9. PAPR Reduction of F-OFDM Using Precoding Techniques

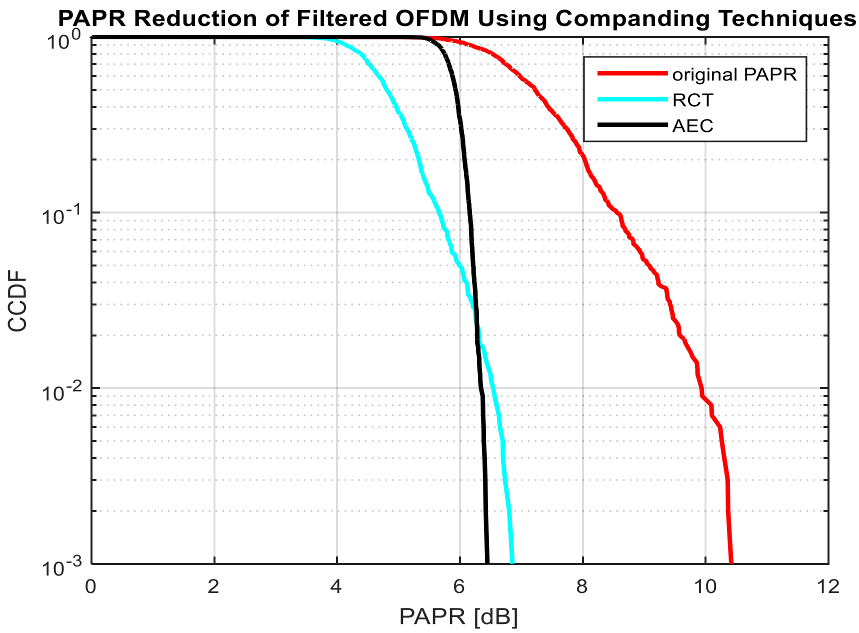

Fig.10. PAPR Reduction of F-OFDM using companding techniques

In figures 11 and 12 we depicted the PAPR reduction graphs for the proposed hybrid techniques Discrete Hartley Transform, Walsh Hadmard Transform (DHT/WHT) precoding techniques with Absolute Exponential companding (AEC) technique, and DHT/WHT precoding with Root Companding (RCT) Technique respectively.

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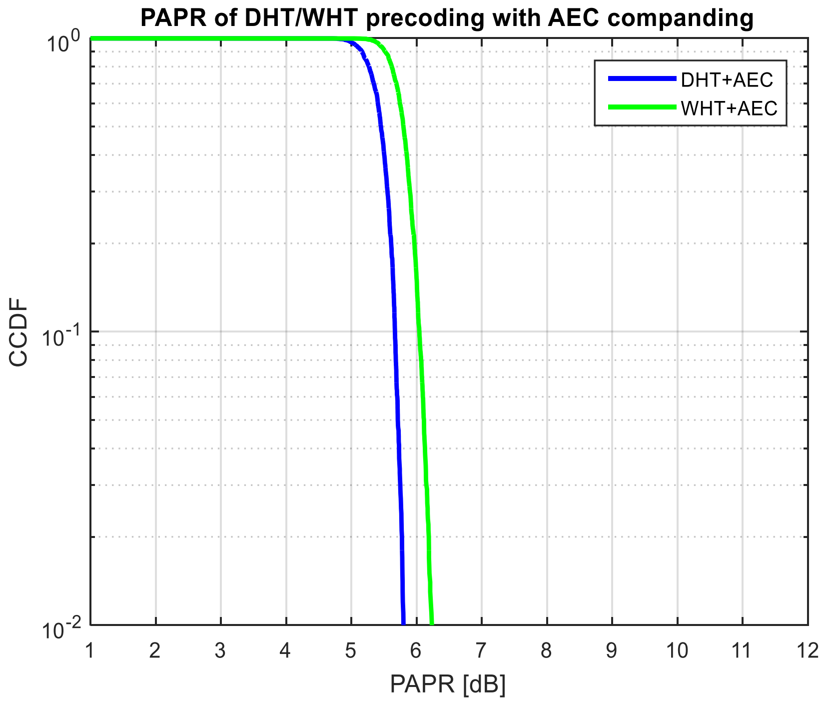

Fig.11.PAPR Reductions of F-OFDM Using First Proposed Hybrid Techniques

In fig. 11 both combined techniques showed an improvement in PAPR reduction; with a 5.8dB for DHT with AEC and 6.2dB for WHT with AEC.Referring to fig. 12the combined DHT/WHT with RCT showed a high reduction in PAPR when compared with the original PAPR of the system. When compared between the two combined techniques the DHT+RCT has shown a better performance than WHT+RCT with a PAPR of 4.7dB and 6dB respectively.

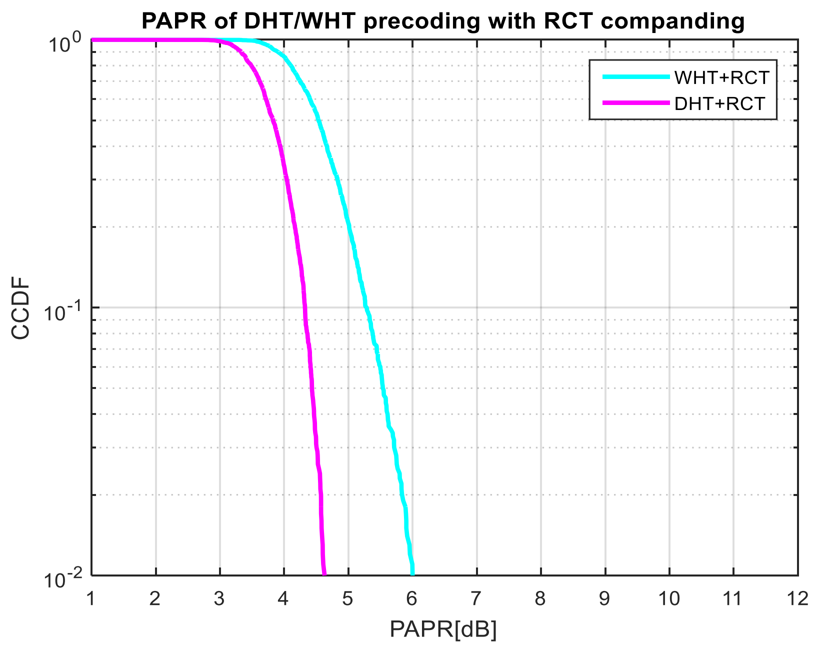

Fig.12. PAPR Reductions of F-OFDM Using Second Propose Hybrid Techniques

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Table 2 and Fig. 13 investigates the BER analysis of the proposed model with existing techniques [19, 20] such as CR, UF-OFDM, and OFDM models. The resultant values demonstrated that the OFDM model has appeared to the worst performer by attaining maximum BER. At the same time, the UF-OFDM and CR models have outperformed the OFDM model. But they failed to exhibit better BER over the proposed model. Among all the methods, the proposed model is found to be superior by achieving a lower BER.

Table 2 Result Anlaysis of Existing Methods with Proposed Method

SNR (dB)	Proposed	CR	UF-OFDM	OFDM
0	0.889531	0.909531	0.959531	0.994531
2	0.886006	0.906006	0.956006	0.991006
4	0.880718	0.900718	0.950718	0.985718
6	0.868380	0.888380	0.938380	0.980380
8	0.852517	0.882517	0.932517	0.982517
10	0.836655	0.866655	0.916655	0.966655
12	0.815504	0.855504	0.905504	0.955504
14	0.774966	0.814966	0.864966	0.924966
16	0.722090	0.782090	0.832090	0.892090
18	0.644539	0.704539	0.764539	0.874539
20	0.528212	0.588212	0.668212	0.778212
22	0.362535	0.442535	0.542535	0.672535
24	0.075243	0.155243	0.255243	0.385243

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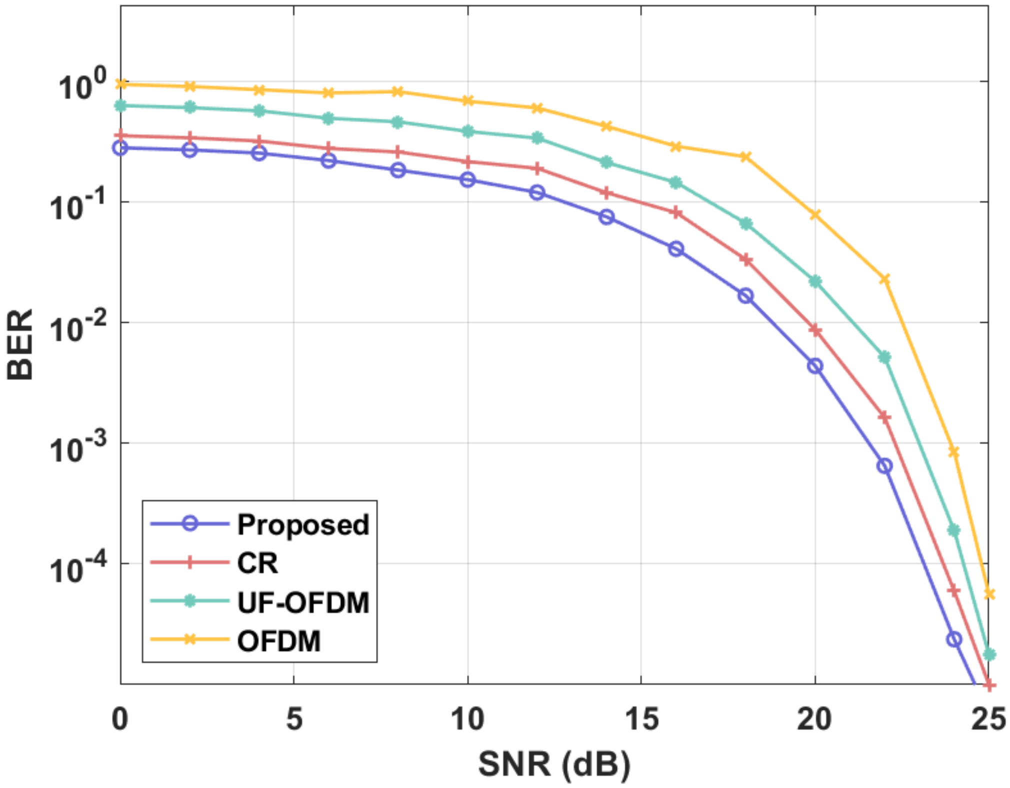

Fig. 13. Comparative BER analysis of proposed with existing models

6. Conclusion

This paper has presented a hybrid technique comprising the precoding with companding for more reduction in the PAPR to maintain the overall system performance. The proposed techniques combine the precoding techniques namely WHT and DHT with two novel non-linear companding techniques such as AEC and RCT. The companding techniques have shown a higher reduction in PAPR when compared with the precoding techniques. The hybrid techniques recorded a higher reduction performance when compared with the conventional single techniques. The DHT+RCT hybrid technique has shown the least PAPR 4.7dB and the highest reduction performance in all used techniques with a reduction of about 6.1 less than that of the original PAPR. At the same time, the WHT+AEC has shown the highest PAPR of 6.2 dB and lowest reduction in all of the proposed hybrid techniques with a reduction of about 4.6 only less than that of the original PAPR.

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